



**COLORADO SCHOOL OF MINES
ELECTRICAL ENGINEERING DEPARTMENT**

EENG 577

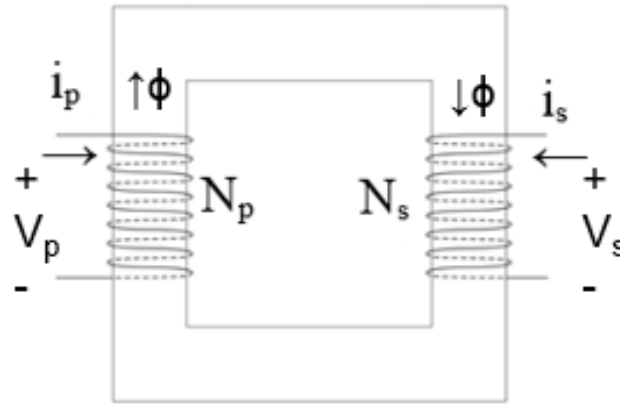
**ADVANCED ELECTRICAL MACHINE DYNAMICS FOR
SMART-GRID SYSTEMS**

**M2-P3 Transformer Equivalent Circuit
Parameters**

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Transformer-Equivalent Circuit

Fig. 1



$$v_j = r_j i_j + \frac{d}{dt} \{\lambda_j\} \quad (1)$$

$$\lambda_j = \sum_{k=1}^n L_{jk} i_k \quad (2)$$

$$v_j = r_j i_j + L_{j1} \frac{di_1}{dt} + \dots + L_{jn} \frac{di_n}{dt} \quad (3)$$

$$L_{pp} = N_p^2 / R,$$

$$L_{ss} = N_s^2 / R, \text{ and}$$

$$L_{ps} = L_{sp} = N_p N_s / R = M \quad (4)$$

$$v_p = r_p \cdot i_p + \frac{d\lambda_p}{dt} \quad (5)$$

$$v_s = r_s \cdot i_s + \frac{d\lambda_s}{dt} \quad (6)$$

$$\lambda_p = L_{pp} \cdot i_p + L_{ps} \cdot i_s \quad (7)$$

$$\lambda_s = L_{ss} \cdot i_s + L_{sp} \cdot i_p \quad (8)$$

Based on the above, and assuming a linear unsaturated magnetic core, and assuming $L_{ps} = L_{sp} = M$, the State Space, SS, model equations for the electromagnetic system considered can be represented as follows:

$$v_p = r_p i_p + L_{pp} \frac{d}{dt} i_p + M \frac{d}{dt} i_s \quad (9)$$

$$v_s = r_s i_s + M \frac{d}{dt} i_p + L_{ss} \frac{d}{dt} i_s \quad (10)$$

Using compact form, the State Space, SS, model equations can be represented as follows:

$$\begin{bmatrix} v_p \\ v_s \end{bmatrix} = \begin{bmatrix} r_p & 0 \\ 0 & r_s \end{bmatrix} \begin{bmatrix} i_p \\ i_s \end{bmatrix} + \begin{bmatrix} L_{pp} & M \\ M & L_{ss} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_p \\ i_s \end{bmatrix} \quad (11)$$

In addition, it could be represented by the following equivalent circuit:

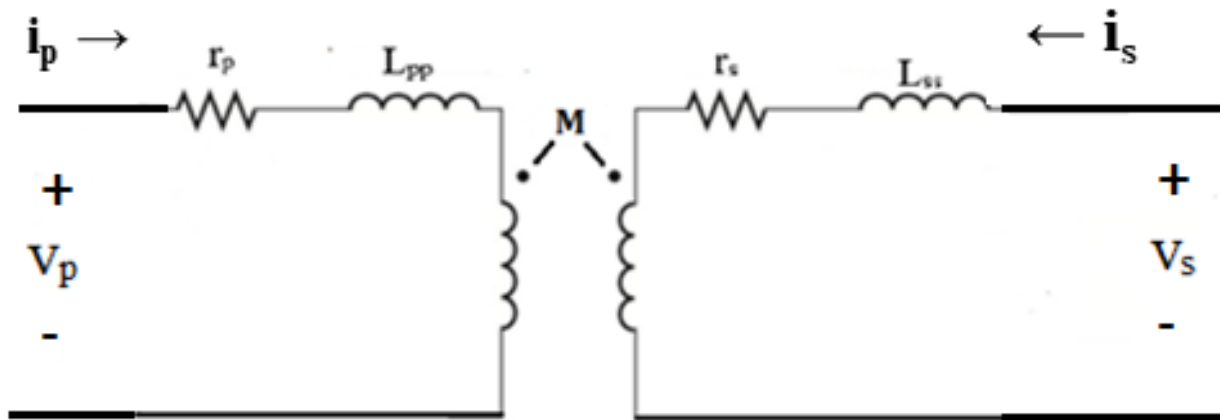


Fig. 2 Transformer State Space Model Equivalent Circuit

Relating SS Model & Transformer Cantilever Circuit Parameters

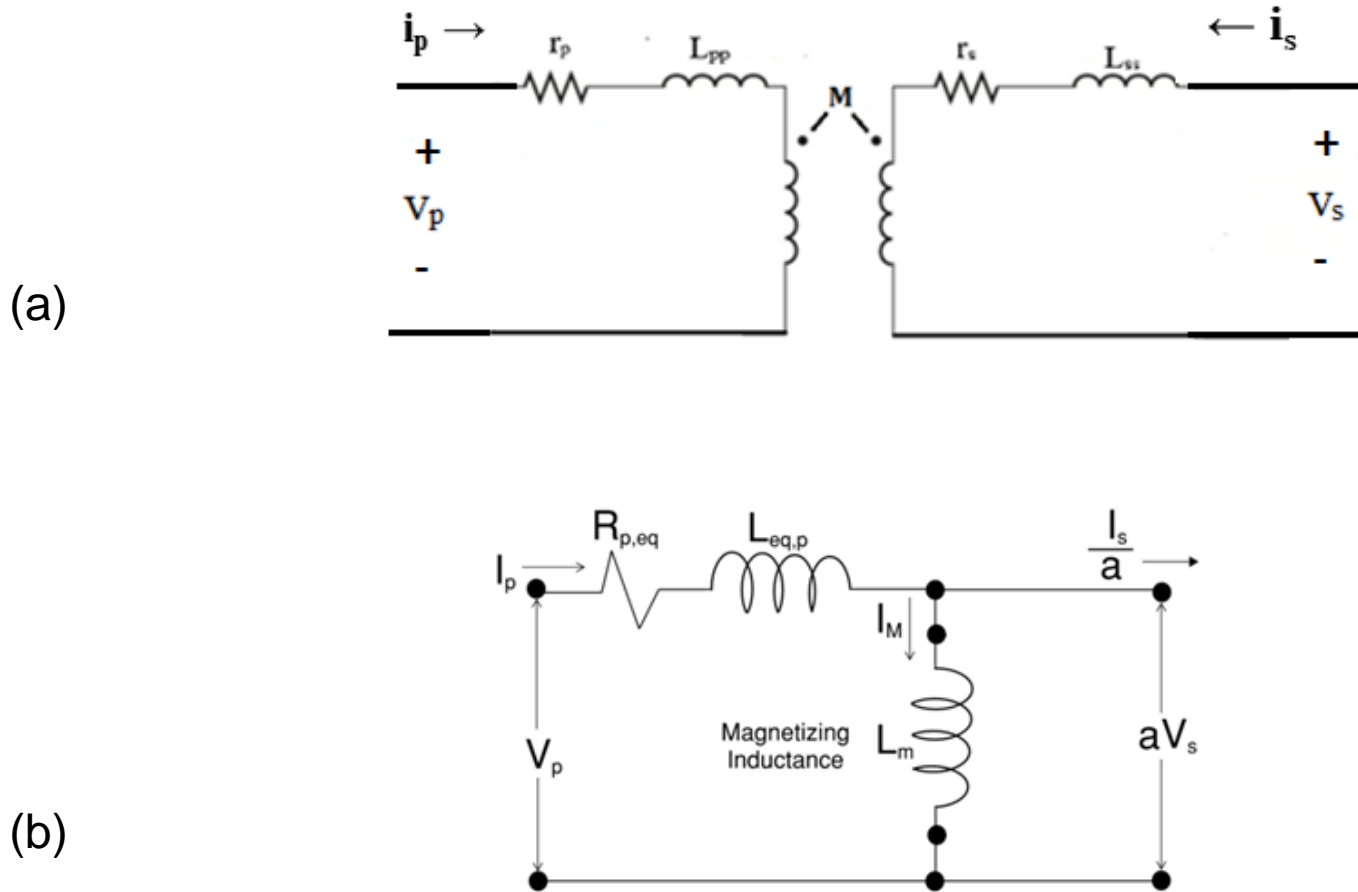


Fig. 3: Transformer Equivalent Circuits:

(a) State Space Model Equivalent Circuit

(b) Simplified Cantilever Equivalent Circuit Referred to Primary Side

Using expression (9) and adding the terms in blue, results in the following :

$$v_p = r_p i_p + L_{pp} \frac{d}{dt} i_p - \frac{N_p}{N_s} M \frac{d}{dt} i_p + \frac{N_p}{N_s} M \frac{d}{dt} i_p + M \frac{d}{dt} i_s \quad (12)$$

Also, collecting similar terms would result in the following:

$$\therefore v_p = r_p i_p + (L_{pp} - \frac{N_p}{N_s} M) \frac{d}{dt} i_p + \frac{N_p}{N_s} M \frac{d}{dt} (i_p + \frac{N_s}{N_p} i_s) \quad (13)$$

Equation (15) can be represented by the circuit of Fig. 4 (a), and can be compared to Fig. 4 (b) of a simplified cantilever equivalent circuit. The two figures relate the self and mutual inductances to the leakage inductance of the cantilever circuit, and thus resulting in the following inductance relationships:

$$L_{eq,p} = L_{lp} + a^2 L_{ls} = L_{pp} - \frac{N_p}{N_s} M \quad \text{and} \quad L_m = \frac{N_p}{N_s} M \quad (14)$$

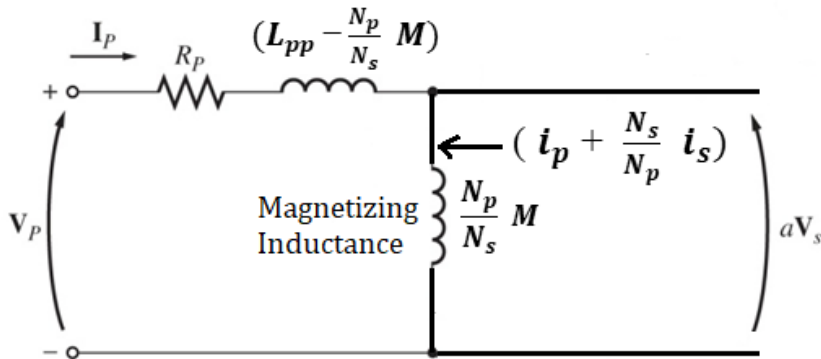
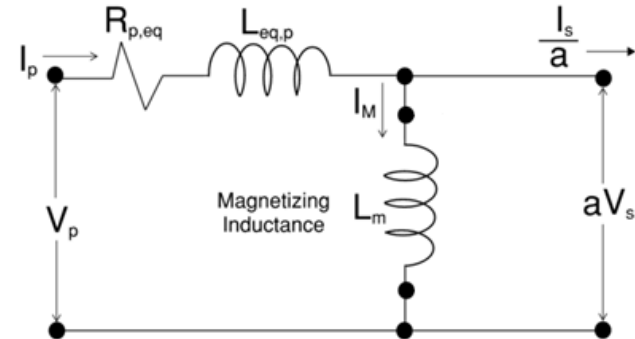


Fig. 4 (a) Transformer SS Model Circuit



(b) Simplified Cantilever Equivalent Circuit